

Signals and Systems Problem-Solving Session

November 3, 2025

DTFS Q1 (§3.6) — Coefficients via symmetry

A discrete-time periodic signal has period $N = 6$ with one period

$$x[n] = \{ 2, 1, 0, -1, -2, 0 \}, \quad n = 0, \dots, 5,$$

extended periodically.

- (i) Decide if $x[n]$ is even/odd/neither (with respect to a suitable center).
- (ii) Compute the DTFS coefficients a_k , $k = 0, \dots, 5$, using any symmetry.

Solution to DTFS Q1 — Detailed

Step 1 (Definition).

$$a_k = \frac{1}{6} \sum_{n=0}^5 x[n] e^{-j \frac{2\pi}{6} kn} = \frac{1}{6} (2e^{-j0} + 1e^{-j \frac{\pi}{3} k} + 0 + (-1)e^{-j\pi k} + (-2)e^{-j \frac{4\pi}{3} k} + 0).$$

Thus

$$a_k = \frac{1}{6} \left(2 + e^{-j \frac{\pi}{3} k} - (-1)^k - 2e^{-j \frac{2\pi}{3} k} \right).$$

Step 2 (Evaluate per k).

$$k = 0 : a_0 = \frac{1}{6} (2 + 1 - 1 - 2) = 0.$$

$$k = 1 : e^{-j\pi/3} = \frac{1}{2} - j \frac{\sqrt{3}}{2}, \quad e^{-j2\pi/3} = -\frac{1}{2} - j \frac{\sqrt{3}}{2}, \quad (-1)^1 = -1.$$

$$a_1 = \frac{1}{6} \left(2 + \frac{1}{2} - j \frac{\sqrt{3}}{2} - (-1) - 2 \left(-\frac{1}{2} - j \frac{\sqrt{3}}{2} \right) \right) = \boxed{\frac{3}{4} - j \frac{\sqrt{3}}{4}}.$$

$$k = 2 : e^{-j2\pi/3} = -\frac{1}{2} - j \frac{\sqrt{3}}{2}, \quad e^{-j4\pi/3} = -\frac{1}{2} + j \frac{\sqrt{3}}{2}, \quad (-1)^2 = 1.$$

$$a_2 = \frac{1}{6} \left(2 - \frac{1}{2} - j \frac{\sqrt{3}}{2} - 1 - 2 \left(-\frac{1}{2} + j \frac{\sqrt{3}}{2} \right) \right) = \boxed{\frac{1}{4} + j \frac{\sqrt{3}}{12}}.$$

DTFS Q2 (§3.7) — Compound operations on coefficients

Let $x[n] \leftrightarrow \{a_k\}$ with period N . Define

$$y[n] = (-1)^n x[n-2] + \frac{1}{2}x[n].$$

Express the DTFS coefficients b_k in terms of $\{a_k\}$.

Solution to DTFS Q2 — Detailed

Step 1 (Break operations). Linearity:

$$b_k = b_k^{(1)} + b_k^{(2)}, \quad y_1[n] = (-1)^n x[n-2], \quad y_2[n] = \frac{1}{2}x[n].$$

Step 2 (Time shift). $x[n - n_0] \Rightarrow a_k e^{-j2\pi kn_0/N}$. For $n_0 = 2$:

$$x[n - 2] \Rightarrow a_k e^{-j4\pi k/N}.$$

Step 3 (Modulation). $(-1)^n = e^{j\pi n}$. For even N , this corresponds to frequency index shift $k \mapsto k - N/2$:

$$e^{j\pi n} x[n] \Rightarrow a_{k-N/2}.$$

Applied after shift:

$$y_1[n] = e^{j\pi n} x[n - 2] \Rightarrow b_k^{(1)} = a_{k-N/2} e^{-j4\pi k/N}.$$

Step 4 (Scaling). $y_2[n] = \frac{1}{2}x[n] \Rightarrow b_k^{(2)} = \frac{1}{2}a_k$.

$$b_k = a_{k-\frac{N}{2}} e^{-j\frac{4\pi}{N}k} + \frac{1}{2} a_k, \quad (\text{indices mod } N).$$

If N is odd, keep modulation as $b_k^{(1)} = \sum_m a_m \delta[k - (m - \frac{N}{2})] e^{-j4\pi k/N}$.

DTFS Q3 (§3.8) — Output of an LTI via harmonics

Given $h[n] = \frac{1}{3}\{1, 1, 1\}$ and

$$x[n] = \cos\left(\frac{\pi n}{2}\right) + 2 \sin\left(\frac{\pi n}{3}\right).$$

(i) Find a fundamental period N . (ii) Compute $y[n] = (h * x)[n]$ using DTFS (harmonic scaling) and simplify in time domain.

Solution to DTFS Q3 — Detailed

Step 1 (Periods). $\Omega_1 = \pi/2 \Rightarrow N_1 = 4$. $\Omega_2 = \pi/3 \Rightarrow N_2 = 6$. LCM $\Rightarrow N = 12$. **Step 2 (Freq response).**

$$H(e^{j\Omega}) = \frac{1}{3}(1 + e^{-j\Omega} + e^{-j2\Omega}) = e^{-j\Omega} \frac{\sin(3\Omega/2)}{3\sin(\Omega/2)}.$$

Step 3 (Evaluate at tones).

$$\Omega = \frac{\pi}{2} : \frac{\sin(3\pi/4)}{3\sin(\pi/4)} = \frac{\frac{\sqrt{2}}{2}}{3 \cdot \frac{\sqrt{2}}{2}} = \frac{1}{3}, \quad H = \frac{1}{3}e^{-j\pi/2} = -\frac{j}{3}.$$

$$\Omega = \frac{\pi}{3} : \frac{\sin(\pi/2)}{3\sin(\pi/6)} = \frac{1}{3 \cdot \frac{1}{2}} = \frac{2}{3}, \quad H = \frac{2}{3}e^{-j\pi/3} = \frac{2}{3}\left(\frac{1}{2} - j\frac{\sqrt{3}}{2}\right).$$

Step 4 (Apply to harmonics).

$$\cos(\Omega_1 n) \Rightarrow \operatorname{Re}\{H(e^{j\Omega_1})e^{j\Omega_1 n}\} = \left(-\frac{j}{3}\right)e^{j\Omega_1 n} \Rightarrow \frac{1}{3}\sin(\Omega_1 n).$$

$$2\sin(\Omega_2 n) \Rightarrow 2\operatorname{Im}\{H(e^{j\Omega_2})e^{j\Omega_2 n}\}.$$

Compute $H(e^{j\Omega_2}) = \frac{1}{3} - j\frac{\sqrt{3}}{3}$. Thus scaling magnitude = $\frac{2}{3}$, phase shift $-\pi/3$. In time:

$$2\sin(\Omega_2 n) \rightarrow \frac{4}{3}\sin(\Omega_2 n).$$

DTFS Q4 (§3.9) — Parseval-based average power

Period $N = 6$, only nonzero

$$a_{\pm 1} = \frac{1}{3}, \quad a_{\pm 2} = \frac{1}{6}, \quad a_0 = a_3 = 0.$$

- (i) Use Parseval to compute average power P_{avg} .
- (ii) Can $x[n]$ be purely real and odd?

Step 1 (Parseval).

$$P_{\text{avg}} = \sum_{k=0}^5 |a_k|^2 = 2 \left(\left| \frac{1}{3} \right|^2 + \left| \frac{1}{6} \right|^2 \right) = 2 \left(\frac{1}{9} + \frac{1}{36} \right) = \frac{5}{18}.$$

Step 2 (Reality & oddness). Real $\Rightarrow a_{-k} = a_k^*$ (true here). Odd $\Rightarrow a_k$ purely imaginary (since time-domain oddness imposes $x[-n] = -x[n]$). Given real $a_{\pm 1}, a_{\pm 2}$, oddness fails. $\Rightarrow$ Real but not purely odd.

CTFT Q5 — Triangular pulse

Let $x(t) = \text{tri}\left(\frac{t}{T}\right)$ (height 1 at $t = 0$, base $2T$). Derive $X(\omega)$ via time-domain convolution or known pairs.

Solution to CTFT Q5 — Detailed

Method (rect * rect). Let $r(t) = \text{rect}(t/T)$. Then $x(t) = (r * r)(t)$.

Fourier domain: $R(\omega) = T \text{sinc}(\frac{\omega T}{2}) \Rightarrow X(\omega) = R^2(\omega)$.

$$X(\omega) = T^2 \text{sinc}^2\left(\frac{\omega T}{2}\right).$$

Sanity check: $x(0) = 1$ and $\frac{1}{2\pi} \int |X(\omega)| d\omega$ finite (energy signal).

CTFT Q6 — Time-multiplication property

Given $x(t) = \text{rect}(t/T)$ with $X(\omega) = T \text{sinc}(\frac{\omega T}{2})$. Find $\mathcal{F}\{t x(t)\}$ using $t x(t) \iff j \frac{d}{d\omega} X(\omega)$.

Solution to CTFT Q6 — Detailed

Step 1. Property: $\mathcal{F}\{t x(t)\} = j \frac{d}{d\omega} X(\omega)$.

$$\frac{d}{d\omega} \left[T \operatorname{sinc}\left(\frac{\omega T}{2}\right) \right] = T \cdot \frac{T}{2} \operatorname{sinc}'\left(\frac{\omega T}{2}\right).$$

Step 2 (Explicit derivative). For $\operatorname{sinc} z = \frac{\sin z}{z}$:

$$\operatorname{sinc}'(z) = \frac{z \cos z - \sin z}{z^2}.$$

Result.

$$\boxed{\mathcal{F}\{t \operatorname{rect}(t/T)\} = j \frac{T^2}{2} \frac{\left(\frac{\omega T}{2}\right) \cos\left(\frac{\omega T}{2}\right) - \sin\left(\frac{\omega T}{2}\right)}{\left(\frac{\omega T}{2}\right)^2}}.$$

CTFT Q7 — Damped sinusoid

For $a > 0$, $x(t) = e^{-at} u(t) \cos(\omega_0 t)$. Find $X(\omega)$.

Solution to CTFT Q7 — Detailed

Step 1 (Split cosine).

$$x(t) = \frac{1}{2}e^{-at}u(t)e^{j\omega_0 t} + \frac{1}{2}e^{-at}u(t)e^{-j\omega_0 t}.$$

Step 2 (Known pair). $\mathcal{F}\{e^{-at}u(t)\} = \frac{1}{a+j\omega}$. Frequency shift gives:

$$\mathcal{F}\{e^{-at}u(t)e^{j\omega_0 t}\} = \frac{1}{a+j(\omega-\omega_0)}.$$

$$\mathcal{F}\{e^{-at}u(t)e^{-j\omega_0 t}\} = \frac{1}{a+j(\omega+\omega_0)}.$$

Combine.

$$X(\omega) = \frac{1}{2} \left(\frac{1}{a+j(\omega-\omega_0)} + \frac{1}{a+j(\omega+\omega_0)} \right).$$

CTFT Q8 — Modulated triangle (AM spectrum)

Let $x(t) = \text{tri}(t/T)$ and $y(t) = x(t) \cos(\omega_0 t)$. Express $Y(\omega)$ and describe its features.

Step 1 (Modulation). $\cos(\omega_0 t) = \frac{1}{2}(e^{j\omega_0 t} + e^{-j\omega_0 t}) \Rightarrow$ spectral shifts:

$$Y(\omega) = \frac{1}{2}X(\omega - \omega_0) + \frac{1}{2}X(\omega + \omega_0).$$

Step 2 (Use Q5). $X(\omega) = T^2 \operatorname{sinc}^2\left(\frac{\omega T}{2}\right).$

$$Y(\omega) = \frac{T^2}{2} \left[\operatorname{sinc}^2\left(\frac{(\omega - \omega_0)T}{2}\right) + \operatorname{sinc}^2\left(\frac{(\omega + \omega_0)T}{2}\right) \right].$$

Two identical sidebands centered at $\pm\omega_0$, each scaled by $1/2$.

CTFT Q9 — Convolution of unequal rectangles

Let $x(t) = \text{rect}(t/T_1)$, $h(t) = \text{rect}(t/T_2)$ with $T_2 > T_1 > 0$. (i) Find $y(t) = (x * h)(t)$. (ii) Find $Y(\omega)$.

Step 1 (Time-domain geometry). Convolution of widths $T_1 < T_2$ gives trapezoid:

$$y(t) = \begin{cases} 0, & |t| \geq \frac{T_1+T_2}{2}, \\ \text{linear rise,} & -\frac{T_1+T_2}{2} < t < -\frac{T_2-T_1}{2}, \\ T_1, & |t| \leq \frac{T_2-T_1}{2}, \\ \text{linear fall,} & \frac{T_2-T_1}{2} < t < \frac{T_1+T_2}{2}. \end{cases}$$

Step 2 (Frequency). $X(\omega) = T_1 \operatorname{sinc}\left(\frac{\omega T_1}{2}\right)$, $H(\omega) = T_2 \operatorname{sinc}\left(\frac{\omega T_2}{2}\right)$.

$$Y(\omega) = X(\omega)H(\omega) = T_1 T_2 \operatorname{sinc}\left(\frac{\omega T_1}{2}\right) \operatorname{sinc}\left(\frac{\omega T_2}{2}\right).$$

CTFT Q10 — RC lowpass response to a pulse

Given $H(\omega) = \frac{1}{1+j\omega RC}$ and $x(t) = \text{rect}(t/T)$. (i) Find $Y(\omega)$. (ii) Obtain $y(t)$ as difference of step responses.

Solution to CTFT Q10 — Detailed

Step 1 (Spectrum). $X(\omega) = T \operatorname{sinc}(\frac{\omega T}{2}) \Rightarrow Y(\omega) = \frac{T \operatorname{sinc}(\omega T/2)}{1+j\omega RC}$.

$$Y(\omega) = \frac{T \operatorname{sinc}(\frac{\omega T}{2})}{1+j\omega RC}.$$

Step 2 (Time-domain construction). Write

$$x(t) = u\left(t + \frac{T}{2}\right) - u\left(t - \frac{T}{2}\right).$$

Causal step response $s(t) = (1 - e^{-t/(RC)})u(t)$. Then

$$y(t) = s\left(t + \frac{T}{2}\right) - s\left(t - \frac{T}{2}\right) = \left(1 - e^{-\frac{t+T/2}{RC}}\right)u\left(t + \frac{T}{2}\right) - \left(1 - e^{-\frac{t-T/2}{RC}}\right)u\left(t - \frac{T}{2}\right).$$

Piecewise flat-rise-hold-fall shape as expected.

CTFT Q11 — Duality & normalization check

Using $\text{rect}(t/T) \iff T \text{sinc}(\omega T/2)$, show that

$$x(t) = \frac{1}{\pi} \text{sinc}(t) \iff X(\omega) = \text{rect}\left(\frac{\omega}{2}\right),$$

and verify $x(0) = \frac{1}{\pi} = \frac{1}{2\pi} \int_{-1}^1 1 d\omega$.

Solution to CTFT Q11 — Detailed

Step 1 (Start from rect–sinc). With $T = 2$:

$$\text{rect}\left(\frac{t}{2}\right) \Longleftrightarrow 2 \text{sinc}\left(\frac{\omega}{1}\right).$$

Step 2 (Duality). Swap $t \leftrightarrow \omega$ and include 2π -convention:

$$\frac{1}{2\pi} \int_{-1}^1 e^{j\omega t} d\omega = \frac{\sin t}{\pi t} = \frac{1}{\pi} \text{sinc}(t) \Longleftrightarrow X(\omega) = \text{rect}\left(\frac{\omega}{2}\right).$$

Step 3 (Normalization at $t = 0$).

$$x(0) = \lim_{t \rightarrow 0} \frac{\sin t}{\pi t} = \frac{1}{\pi} = \frac{1}{2\pi} \int_{-1}^1 1 d\omega.$$