

# EEE 321: Signals and Systems

## The Discrete-Time Fourier Transform

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# Outline

- 1 Cheat Sheet: DTFT Formulas
- 2 Q1: Basic Transform
- 3 Q2: DTFT Properties
- 4 Q3: Inverse DTFT
- 5 Q4: LCCDE & System Response

## Recall: DTFT Formulas

The Discrete-Time Fourier Transform (DTFT) pair is defined as follows:

### Analysis Equation (Transform)

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

### Synthesis Equation (Inverse Transform)

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega})e^{j\omega n}d\omega$$

**Important Note:**  $X(e^{j\omega})$  is periodic with respect to  $\omega$  with period  $2\pi$ .

## Question 1: Basic DTFT Calculation

**Problem:** Determine the DTFT of the following causal signal:

$$x[n] = a^n u[n], \quad |a| < 1$$

Where  $u[n]$  is the unit step function.

# Question 1: Basic DTFT Calculation

**Problem:** Determine the DTFT of the following causal signal:

$$x[n] = a^n u[n], \quad |a| < 1$$

Where  $u[n]$  is the unit step function.

**Solution:** Using the definition equation:

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} a^n u[n] e^{-j\omega n}$$

Since  $u[n]$  is non-zero only for  $n \geq 0$ :

$$X(e^{j\omega}) = \sum_{n=0}^{\infty} a^n e^{-j\omega n} = \sum_{n=0}^{\infty} (ae^{-j\omega})^n$$

## Question 1: Solution Continued

This is an infinite geometric series:

$$\sum_{k=0}^{\infty} r^k = \frac{1}{1-r}, \quad \text{if } |r| < 1$$

Here,  $r = ae^{-j\omega}$ . The convergence condition implies  $|ae^{-j\omega}| = |a| < 1$ , which is given.

### Result

$$X(e^{j\omega}) = \frac{1}{1 - ae^{-j\omega}}$$

This is a fundamental pair often used in system analysis.

## Question 2: Time Shifting and Linearity

**Problem:** Find the DTFT of the following signal using properties (avoid direct summation).

$$x[n] = \left(\frac{1}{2}\right)^{n-1} u[n-1]$$

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**Strategy:** Instead of calculating the sum from scratch, we use the result from Q1 and the *Time Shifting Property*.

$$x[n - n_0] \longleftrightarrow e^{-j\omega n_0} X(e^{j\omega})$$



## Question 2: Solution

Let us define a standard signal  $y[n] = (1/2)^n u[n]$ .

From Question 1, we know (with  $a = 1/2$ ):

$$Y(e^{j\omega}) = \frac{1}{1 - \frac{1}{2}e^{-j\omega}}$$

The given signal  $x[n]$  is simply  $y[n]$  shifted to the right by 1 unit:

$$x[n] = y[n - 1]$$

Applying the Time Shifting property ( $n_0 = 1$ ):

$$X(e^{j\omega}) = e^{-j\omega \cdot 1} Y(e^{j\omega})$$

### Result

$$X(e^{j\omega}) = \frac{e^{-j\omega}}{1 - \frac{1}{2}e^{-j\omega}}$$

## Question 3: Inverse DTFT

**Problem:** Determine the signal  $x[n]$  corresponding to the following frequency spectrum (periodic with  $2\pi$ ):

$$X(e^{j\omega}) = 2\pi\delta(\omega - \omega_0), \quad |\omega| < \pi$$

## Question 3: Inverse DTFT

**Problem:** Determine the signal  $x[n]$  corresponding to the following frequency spectrum (periodic with  $2\pi$ ):

$$X(e^{j\omega}) = 2\pi\delta(\omega - \omega_0), \quad |\omega| < \pi$$

**Solution:** Use the Synthesis Equation (Inverse Transform).

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} 2\pi\delta(\omega - \omega_0) e^{j\omega n} d\omega$$

## Question 3: Solution Continued

The Dirac delta function inside the integral utilizes the *Sifting Property*, evaluating the function only at  $\omega = \omega_0$ .

$$x[n] = \int_{-\pi}^{\pi} \delta(\omega - \omega_0) e^{j\omega n} d\omega$$

$$x[n] = e^{j\omega_0 n}$$

### Interpretation

This result demonstrates the duality: A complex exponential in the time domain ( $e^{j\omega_0 n}$ ) corresponds to a single impulse in the frequency domain. This is a key concept in Fourier analysis.

## Question 4: Difference Equations

**Problem:** An LTI system is characterized by the following Linear Constant-Coefficient Difference Equation (LCCDE):

$$y[n] - \frac{3}{4}y[n-1] + \frac{1}{8}y[n-2] = x[n]$$

- a) Find the Frequency Response,  $H(e^{j\omega})$ .
- b) Find the Impulse Response,  $h[n]$ .

## Question 4: Difference Equations

**Problem:** An LTI system is characterized by the following Linear Constant-Coefficient Difference Equation (LCCDE):

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- a) Find the Frequency Response,  $H(e^{j\omega})$ .
- b) Find the Impulse Response,  $h[n]$ .

**Solution (a):** Take the DTFT of both sides (using Linearity and Time Shifting):

$$Y(e^{j\omega}) - \frac{3}{4}e^{-j\omega}Y(e^{j\omega}) + \frac{1}{8}e^{-j2\omega}Y(e^{j\omega}) = X(e^{j\omega})$$

## Question 4: Frequency Response

Factor out  $Y(e^{j\omega})$ :

$$Y(e^{j\omega}) \left[ 1 - \frac{3}{4}e^{-j\omega} + \frac{1}{8}e^{-j2\omega} \right] = X(e^{j\omega})$$

Since the frequency response is defined as  $H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})}$ :

$$H(e^{j\omega}) = \frac{1}{1 - \frac{3}{4}e^{-j\omega} + \frac{1}{8}e^{-j2\omega}}$$

## Question 4: Impulse Response (b)

To find  $h[n]$ , we need the inverse DTFT of  $H(e^{j\omega})$ . We factor the denominator. Let  $z^{-1} = e^{-j\omega}$  for simplicity.

$$1 - \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2} = \left(1 - \frac{1}{2}z^{-1}\right) \left(1 - \frac{1}{4}z^{-1}\right)$$

Using Partial Fraction Expansion:

$$H(e^{j\omega}) = \frac{A}{1 - \frac{1}{2}e^{-j\omega}} + \frac{B}{1 - \frac{1}{4}e^{-j\omega}}$$

Solving for coefficients yields:  $A = 2$ ,  $B = -1$ .



## Question 4: Final Result

$$H(e^{j\omega}) = \frac{2}{1 - \frac{1}{2}e^{-j\omega}} - \frac{1}{1 - \frac{1}{4}e^{-j\omega}}$$

Recall the transform pair from Question 1:  $\frac{1}{1 - ae^{-j\omega}} \leftrightarrow a^n u[n]$ .  
Applying the inverse transform term by term:

### Result

$$h[n] = 2 \left(\frac{1}{2}\right)^n u[n] - \left(\frac{1}{4}\right)^n u[n]$$

# Summary & Next Steps

## Key Takeaways:

- Analyzing signals using the DTFT analysis equation.
- Simplifying calculations using the Time Shifting property.
- Deriving the Frequency Response from LCCDEs.
- Calculating Inverse DTFT via Partial Fraction Expansion.

**Recommended Reading:** Review Section 5.4 (The Convolution Property) and Parseval's Relation (Energy Conservation) in the textbook.

# EEE 321: Signals and Systems

## The Discrete-Time Fourier Transform - 2

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## Question 5: Problem Statement

Consider a filter whose frequency response  $H(e^{j\Omega})$  is specified over the frequency interval  $0 \leq \Omega < 2\pi$  as follows:

$$H(e^{j\Omega}) = \begin{cases} 5, & \frac{3\pi}{4} \leq \Omega \leq \frac{5\pi}{4} \\ 0, & \text{otherwise} \end{cases}$$

- i. Find and plot  $H(e^{j\Omega})$  in the interval  $(-\pi, \pi]$ . Specify the type of this filter (e.g. low-pass, high-pass, band-pass, band-stop).
- ii. Calculate the output signal  $y[n]$  of this filter when the input signal is given as:

$$x[n] = 10 + \cos(88\pi n) + 5 \sin(11\pi n) - 5 \cos\left(\frac{23\pi n}{8}\right)$$

- iii. Calculate the output signal  $y[n]$  of this filter when the input signal is given as:

$$x[n] = 2(-1)^n$$

## Question 5(i): Filter Identification

Find and plot  $H(e^{j\Omega})$  in the interval  $(-\pi, \pi]$ . Specify the type of this filter.

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Find and plot  $H(e^{j\Omega})$  in the interval  $(-\pi, \pi]$ . Specify the type of this filter.

**Answer:** Range  $[\frac{3\pi}{4}, \frac{5\pi}{4}]$  maps to high frequencies near  $\pi$  (due to  $2\pi$  periodicity).

- The active region in  $(-\pi, \pi]$  is  $[-\pi, -\frac{3\pi}{4}] \cup [\frac{3\pi}{4}, \pi]$ .
- Filter Type: **High-Pass Filter**.

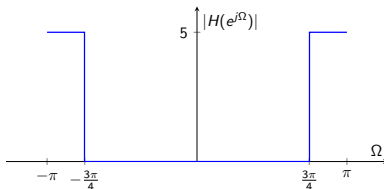
## Question 5(i): Filter Identification

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**Answer:** Range  $[\frac{3\pi}{4}, \frac{5\pi}{4}]$  maps to high frequencies near  $\pi$  (due to  $2\pi$  periodicity).

- The active region in  $(-\pi, \pi]$  is  $[-\pi, -\frac{3\pi}{4}] \cup [\frac{3\pi}{4}, \pi]$ .
- Filter Type: **High-Pass Filter**.

**Visual Solution:**



## Question 5(ii): Output Calculation 1

**Problem:** Calculate the output  $y[n]$  for input:

$$x[n] = 10 + \cos(88\pi n) + 5 \sin(11\pi n) - 5 \cos\left(\frac{23\pi n}{8}\right)$$



## Question 5(ii): Output Calculation 1

**Problem:** Calculate the output  $y[n]$  for input:

$$x[n] = 10 + \cos(88\pi n) + 5 \sin(11\pi n) - 5 \cos\left(\frac{23\pi n}{8}\right)$$

**Solution:**

- ① **DC (10):**  $\Omega = 0$  (Stopband)  $\rightarrow 0$ .
- ②  **$\cos(88\pi n)$ :**  $\Omega \equiv 0$  (Stopband)  $\rightarrow 0$ .
- ③  **$5 \sin(11\pi n)$ :**  $\Omega \equiv \pi$  (Passband). Gain = 5.
- ④  **$-5 \cos(\frac{23\pi n}{8})$ :**  $\Omega = \frac{7\pi}{8}$ . Since  $\frac{7\pi}{8} > \frac{6\pi}{8}$ , it is in Passband. Gain = 5.

$$y[n] = 25 \sin(11\pi n) - 25 \cos\left(\frac{23\pi n}{8}\right)$$

## Question 5(iii): Output Calculation 2

**Problem:** Calculate the output  $y[n]$  for input:

$$x[n] = 2(-1)^n$$

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**Problem:** Calculate the output  $y[n]$  for input:

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**Solution:**

$$x[n] = 2 \cos(\pi n) \implies \Omega = \pi$$

$\pi$  is in the center of the passband ( $\frac{3\pi}{4} \leq \pi \leq \frac{5\pi}{4}$ ). The gain is 5.

$$y[n] = 5 \cdot x[n] = 10(-1)^n$$

## Question 6: Problem Statement

Let  $H(e^{j\Omega})$  be the frequency response of an LTI system. It is given that:

- The system is **causal**.
- The impulse response  $h[n]$  of the system is **real-valued**.
- $DTFT(h[n] - h[-n]) = 4j \sin(2\Omega) - 2j \sin(3\Omega) + 10j \sin(5\Omega)$ .
- $\int_{4\pi}^{6\pi} H(e^{j\Omega}) H^*(e^{j\Omega}) d\Omega = 64\pi$ .
- $h[0] > 0$ .

**Task:** Evaluate and plot  $h[n]$ .

## Question 6: Step 1 - Time Domain Analysis

$$\mathcal{F}\{h[n] - h[-n]\} = 4j \sin(2\Omega) - 2j \sin(3\Omega) + 10j \sin(5\Omega)$$

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$$\mathcal{F}\{h[n] - h[-n]\} = 4j \sin(2\Omega) - 2j \sin(3\Omega) + 10j \sin(5\Omega)$$

Using  $2j \sin(k\Omega) \leftrightarrow \delta[n+k] - \delta[n-k]$ :

$$h[n] - h[-n] = 2(\delta[n+2] - \dots) - (\delta[n+3] - \dots) + 5(\delta[n+5] - \dots)$$

Since causal ( $h[-k] = 0$  for  $k > 0$ ):

$$h[2] = -2, \quad h[3] = 1, \quad h[5] = -5$$

## Question 6: Step 2 - Parseval's Relation

Condition:  $\int_{2\pi} |H(e^{j\Omega})|^2 d\Omega = 64\pi.$

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Condition:  $\int_{2\pi} |H(e^{j\Omega})|^2 d\Omega = 64\pi$ .

$$64\pi = 2\pi \sum |h[n]|^2 \implies 32 = \sum |h[n]|^2$$

$$32 = h[0]^2 + (-2)^2 + (1)^2 + (-5)^2$$

$$32 = h[0]^2 + 30 \implies h[0] = \sqrt{2}$$



## Question 6: Final Result

$$h[n] = \sqrt{2}\delta[n] - 2\delta[n-2] + \delta[n-3] - 5\delta[n-5]$$

